



# Electronic Circuit Analysis

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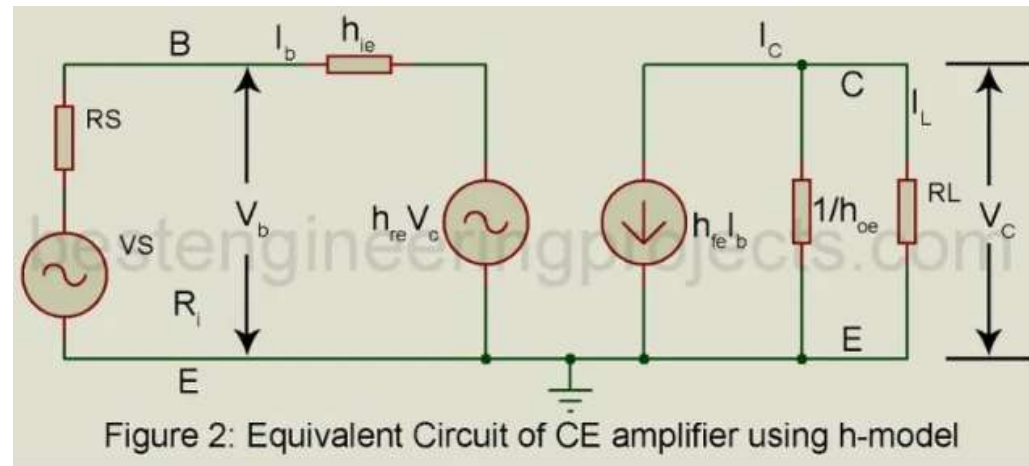
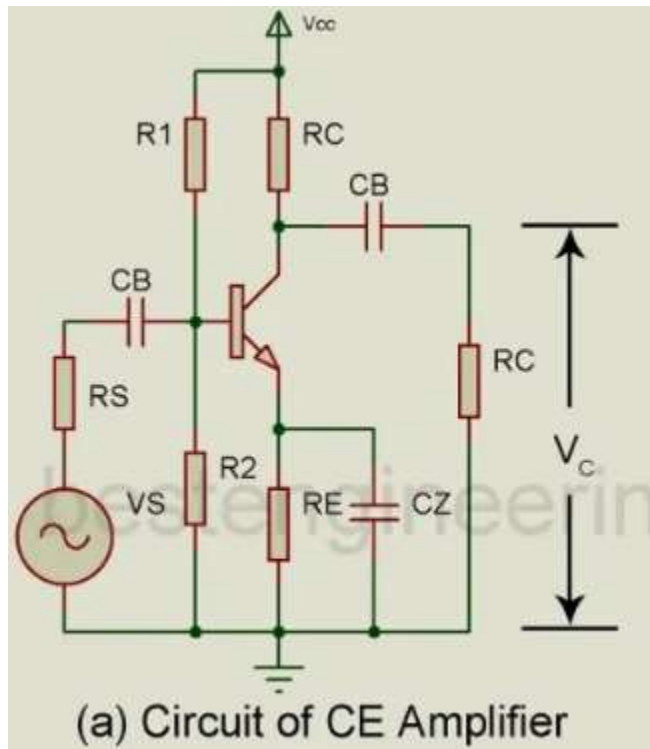
Bapatla Engineering College

## UNIT – I

**BJT at low frequency:** Transistor Hybrid model, Determination of h-parameters from characteristics, Analysis of transistor amplifier using h-parameter model, Emitter follower, Millers theorem and its dual, Cascading transistor amplifiers, Simplified CE & CC Hybrid models, CE Amplifier with an Emitter Resistance, High Input Resistance Transistor Circuits – Darlington pair, Boot strapped Darlington pair.

**FET at low frequency:** FET Small signal model, Common Source and Common Drain configurations at low frequencies.

# Analysis of Common Emitter Amplifier using H Parameter



For analysis, we replace the transistor with its small-signal two generator h-parameter model. This results in the equivalent circuit of Figure 2. We assume sinusoidal input.

## Current Gain or Current Amplification:

Current gain is defined as the ratio of the load current  $I_L$  to the input current  $I_b$ . Thus,

$$\text{Current Gain } A_I = \frac{I_L}{I_b} = -\frac{I_c}{I_b} \quad \dots\dots(1)$$

$$\text{But from figure 2, } I_c = h_{fe} \times I_b + h_{oe} \times V_c \quad \dots\dots(2)$$

$$\text{Also } V_c = I_L \times R_L = -I_c \times R_L \quad \dots\dots(3)$$

Combining Equation (2) and (3) we get,

$$I_c = h_{fe} I_b - h_{oe} \times I_c \times R_L \text{ or } (1 + h_{oe} \times R_L) I_c = h_{fe} \times I_b$$

$$\text{Hence current gain } A_I = -\frac{I_c}{I_b} = -\frac{h_{fe}}{1 + h_{oe} \times R_L} \quad \dots\dots(4)$$

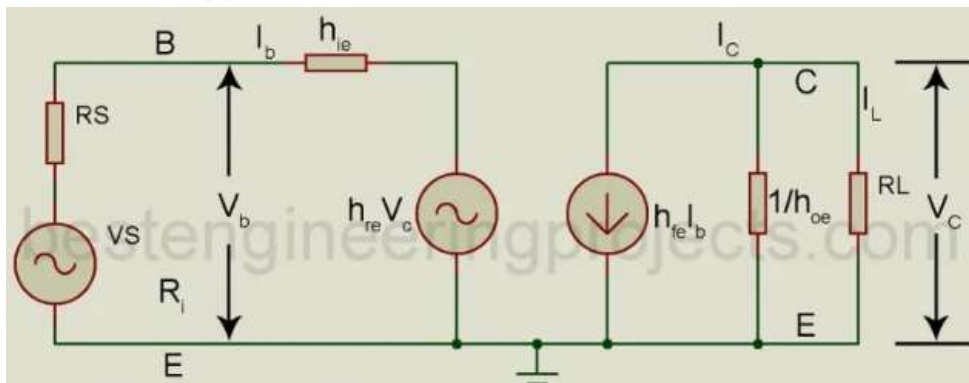


Figure 2: Equivalent Circuit of CE amplifier using h-model

## Input Impedance $R_i$ :

This is the impedance between the input terminals B and E looking into the amplifier as shown in Figure 2 and is, therefore, given by,

$$R_i = \frac{V_b}{I_b} \quad \dots\dots(5)$$

$$\text{From figure 2 } V_b = h_{ie} \times I_b + h_{re} \times V_c \quad \dots\dots(6)$$

$$\text{But } V_c = -I_c \times R_L = A_I I_b R_L \quad \dots\dots(7)$$

Substituting the value of  $V_c$  from Equation (7) into Equation (6) we get,

$$V_b = h_{ie} \times I_b + h_{re} A_I I_b R_L \quad \dots\dots(8)$$

$$\text{Hence input impedance } R_i = \frac{V_b}{I_b} = h_{ie} + h_{re} A_I R_L \quad \dots\dots(9)$$

$$= h_{ie} - \frac{h_{fe} h_{re}}{h_{oe} + Y_L} \quad \dots\dots(10)$$

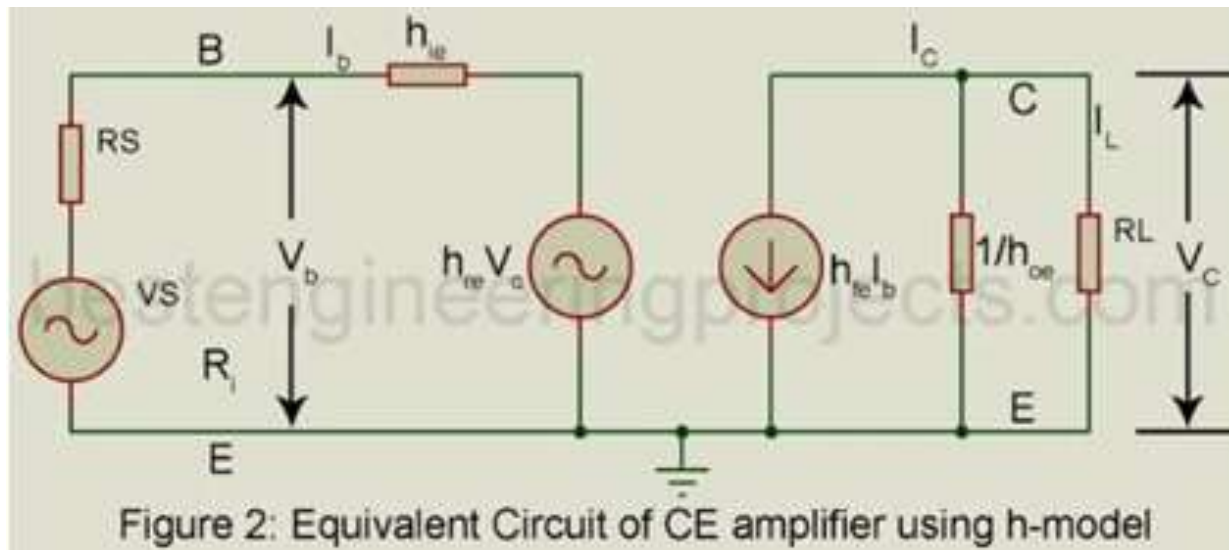
$$\text{Where } Y_L = \frac{1}{R_L}$$

From Equation (10) we find that the input impedance  $R_i$  is also a function of load resistance  $R_L$ .

## Voltage Gain or Voltage Amplification | Analysis of Common Emitter Amplifier using H Parameter:

It is the ratio of the output voltage  $V_c$  to the input voltage  $V_b$ . Thus,

$$\text{Voltage Gain } A_v = \frac{V_c}{V_b} = -\frac{I_c R_L}{I_b R_i} = \frac{A_I R_L}{R_i} \quad \dots(11)$$



## Output Admittance $Y_0$ :

It is the ratio of the output current  $I_c$  to the output voltage  $V_c$  with  $V_s = 0$ . Hence

$$Y_0 = \frac{I_c}{V_c} \text{ with } V_s = 0 \quad \dots\dots(12)$$

On substituting the value of  $I_c$  from Equation (2) into Equation (12) we get,

$$Y_0 = h_{fe} \times \frac{I_b}{V_c} + h_{oe} \quad \dots\dots(13)$$

But with  $V_s = 0$ , Figure 2 gives  $(R_s + h_{ie}) I_b + h_{re} V_c = 0$

$$\text{Or } \frac{I_b}{V_c} = - \frac{h_{re}}{h_{ie} + R_s} \quad \dots\dots(14)$$

Combining Equation (13) and (14) we get,  $Y_0 = h_{oe} - \frac{h_{fe} \times h_{re}}{h_{ie} + R_s} \quad \dots\dots(15)$

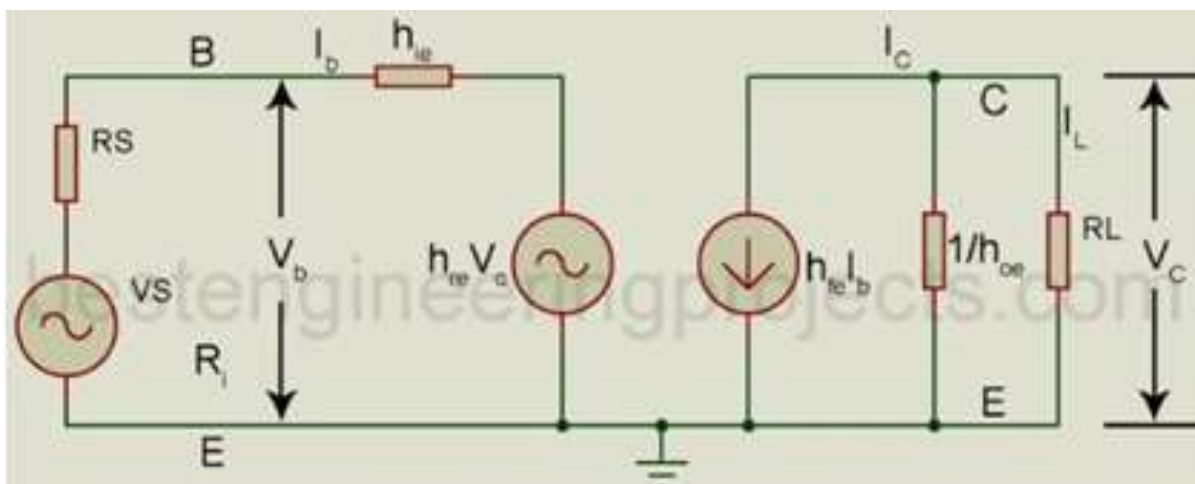


Figure 2: Equivalent Circuit of CE amplifier using h-model

Equation (15) shows that the output admittance  $Y_0$  is a function of source resistance  $R_s$ . If the source impedance is purely resistive, then the output impedance  $Z_0$  is real i.e. purely conductive.

Output impedance  $R_0 = \frac{1}{Y_0}$  ....(16)

In the calculation of  $Y_0$ ,  $R_L$  has been considered external to the amplifier. If we include  $R_L$  in parallel with  $R_0$ , we get the output terminal impedance  $Z_t$  given by,

$$Z_{t} = \frac{R_0 R_L}{R_0 + R_L} \quad \text{.....(17)}$$

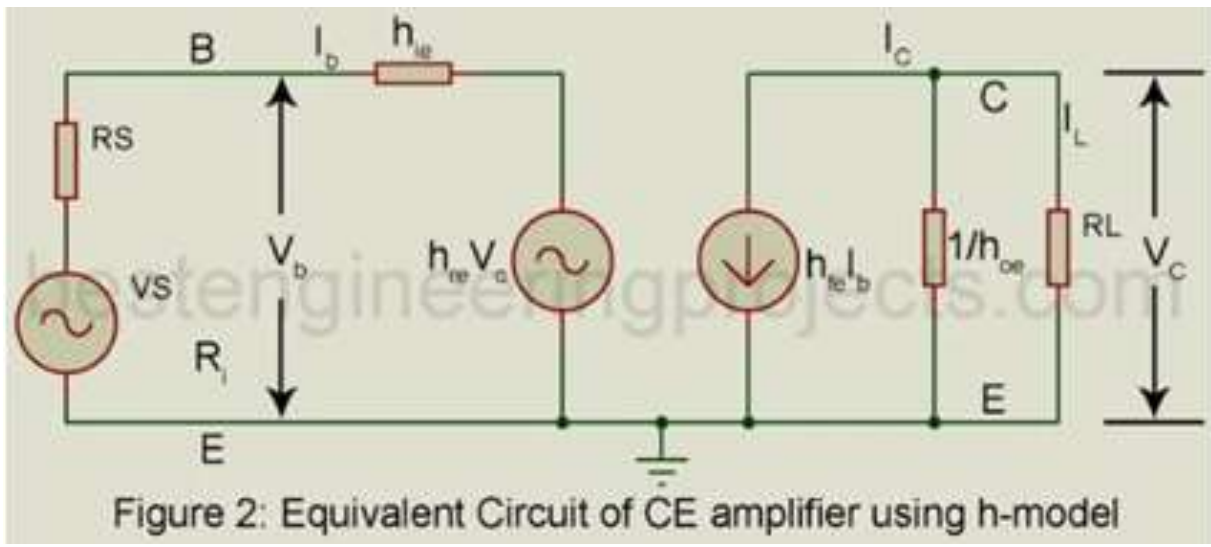


Figure 2: Equivalent Circuit of CE amplifier using h-model



## Overall Voltage Gain Considering $R_s$ :

Source voltage  $V_s$  applied at the input of an amplifier results in voltage  $V_b$  between base and emitter terminals (input terminals) of the transistor and voltage  $V_c$  at the output. Then the overall voltage gain considering the source resistance is given by

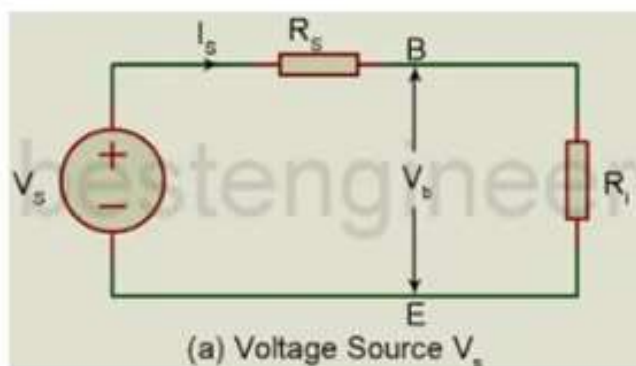
$$A_{V_S} = \frac{V_c}{V_s} = \frac{V_c}{V_b} \times \frac{V_b}{V_s} = A_V \frac{V_b}{V_s} \quad \dots\dots(18)$$

Figure 3(a) given the driven voltage source  $V_s$  with source resistance  $R_s$  in series. This form of an equivalent circuit for the energy source is known as Thevenin's equivalent source. This energy source then drives the amplifier represented by its input resistance  $R_i$ .

Then 
$$V_b = V_s \times \frac{R_i}{R_i + R_s} \quad \dots\dots(19)$$

Hence overall voltage gain 
$$A_{V_S} = A_V \times \frac{R_i}{R_i + R_s} \quad \dots\dots(20)$$

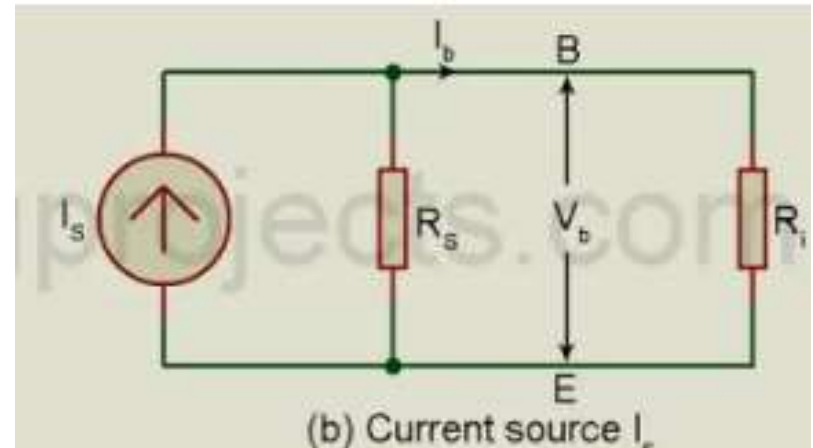
If  $R_s = 0$ , then  $A_{V_S} = A_V$ . Thus,  $A_V$  forms a special case of  $A_{V_S}$  with  $R_s = 0$ .



## Overall Current Gain Considering $R_s$ :

We may replace the voltage source  $V_s$  with series source resistance  $R_s$  by what is known as Norton's equivalent source shown in Figure 3(b), consisting of current source  $I_s$  with source resistance  $R_s$  in the shunt. This current source drives the amplifier resulting in  $I_b$  at the input terminals of the amplifier and current  $I_L$  through the load impedance. Then the overall current gain  $A_{IS}$  is given by:

$$A_{IS} = \frac{I_L}{I_s} = \frac{-I_C}{I_b} \times \frac{I_b}{I_s} = A_I \times \frac{I_b}{I_s} \quad \dots(21)$$



From Figure 3(b), 
$$I_b = I_s \frac{R_s}{R_s + R_i} \quad \dots(22)$$

Hence overall current gain 
$$A_{IS} = A_I \frac{R_s}{R_s + R_i} \quad \dots(23)$$

From Equations (20) and Equation (23) we get, 
$$A_{VS} = A_{IS} \times \frac{R_i}{R_s} \quad \dots(24)$$

Equation (24) is true provided that the voltage source  $V_s$  and the current source  $I_s$  have the same source resistance  $R_s$ .

**Table 1: Result of small signal analysis of low frequency ce amplifier**

$$A_I = -\frac{h_{fe}}{1 + h_{oe} \times R_L}$$

$$A_V = \frac{A_I \times R_L}{R_i}$$

$$R_i = h_{ie} + h_{re} \times A_I \times R_L$$

$$A_{VS} = \frac{A_V R_i}{R_i + R_s}$$

$$Y_0 = h_{oe} - \frac{h_{re} h_{fe}}{h_{ie} + R_s} = \frac{1}{Z_0}$$

$$A_{IS} = \frac{A_I R_s}{R_i + R_s}$$

Ex:1. The transistor is connected as CE amplifier with h-parameters  
 $h_{ie} = 1100 \Omega$ ,  $h_{re} = 2.5 \times 10^{-4}$ ,  $h_{fe} = 50$  and  $h_{oe} = 24 \mu A/V$ . If  
 $R_L = 10 K\Omega$  and  $R_S = 1 K\Omega$ , find the various gains and  
the i/p and o/p impedances.

Solution:  $A_I = -\frac{h_{fe}}{1 + h_{oe} R_L} = -\frac{50}{1 + 25 \times 10^{-6} \times 10^4} = -40.0$

$(Z_i = R_i) = h_{ie} + h_{re} A_I R_L = 1100 - 2.5 \times 10^{-4} \times 40 \times 10^4 = 1000 \Omega$

$A_V = \frac{A_I R_L}{R_i} = \frac{(-40) \times 10 \times 10^3}{1000} = -400$

$A_{Vs} = \frac{A_V R_i}{R_i + R_S} = \frac{(-400)(1000)}{1000 + 1000} = -200$

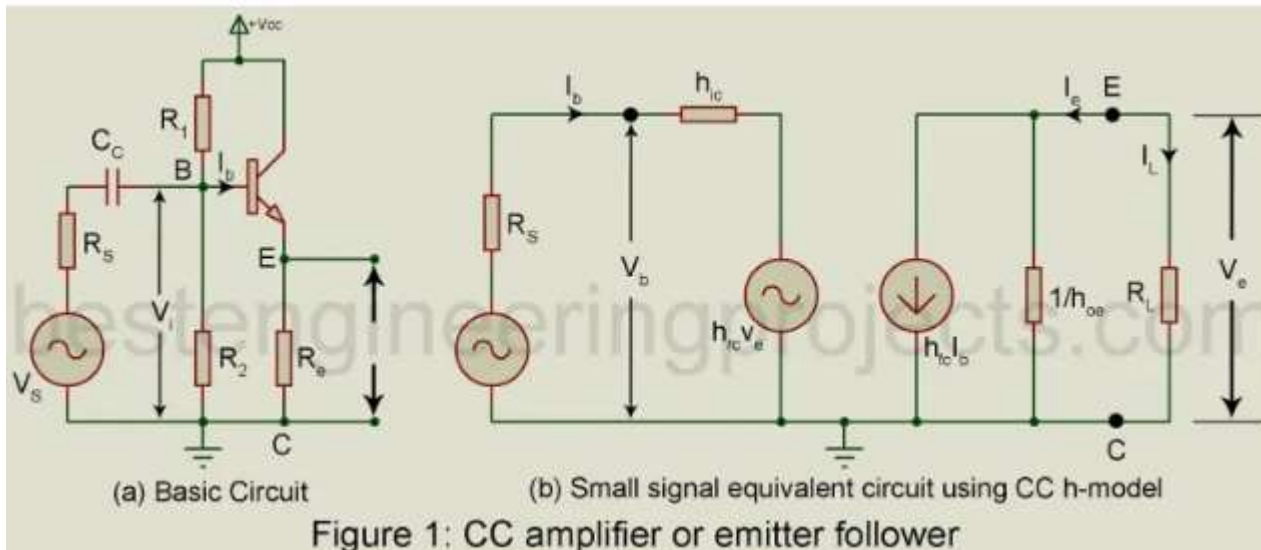
$A_{Is} = \frac{A_I R_S}{R_i + R_S} = \frac{(-40)(1000)}{1000 + 1000} = -20$

$Y_o = h_{oe} - \frac{h_{fe} h_{re}}{h_{ie} + R_S} = 25 \times 10^{-6} - \frac{50 \times 2.5 \times 10^{-4}}{1100 + 1000} = 19 \times 10^{-6} S$   
 $= 19 \mu A/V$

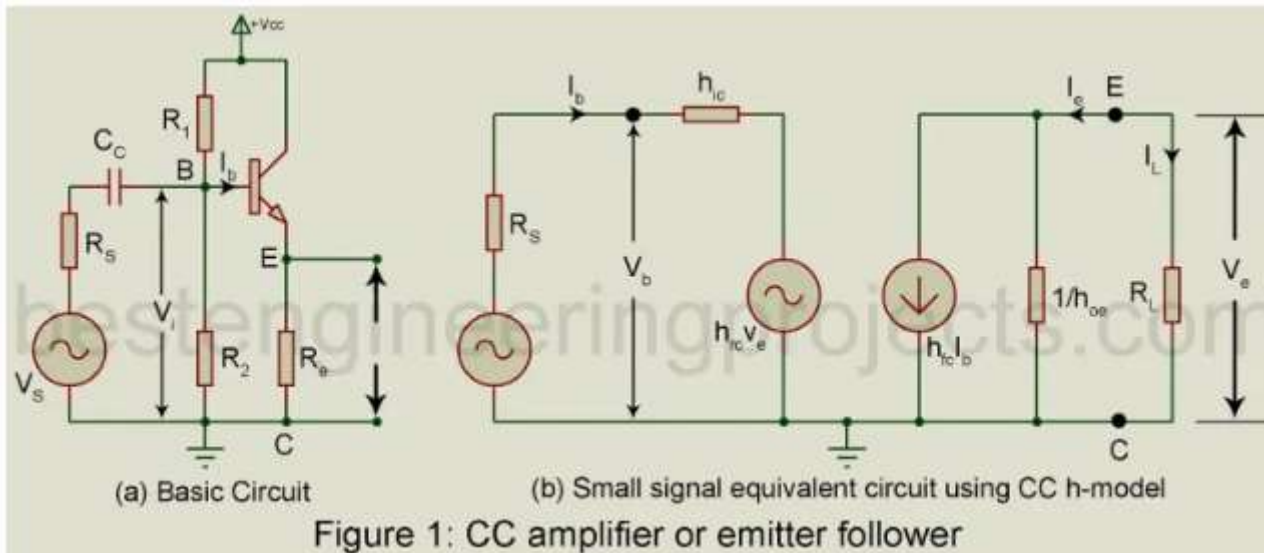
$Z_o = \frac{1}{Y_o} = \frac{1}{19 \times 10^{-6}} = \frac{10^6}{19} \Omega = 52.6 K\Omega$

# Analysis of EMITTER FOLLOWER using h-parameter model

**Common Collector Amplifier or the Emitter Follower:** Figure 1(a) gives the basic circuit of a common collector amplifier, popularly called emitter follower. Its voltage gain is close to unity (one) and, therefore, any increment in the input voltage i.e. the base voltage appears as an equal increment in the output voltage across the load resistor in the emitter circuit. Thus, the emitter may be said to follow the input signal and hence the name emitter follower.



The equivalent circuit of figure 1(a) is similar to that of CE amplifier of Figure 2. Hence the analysis procedure is exactly the same. The expression of  $A_i$ ,  $R_i$ ,  $A_v$ ,  $A_{iS}$ ,  $A_{vS}$  and  $Y_o$  are, therefore, the same as of CE amplifier except that h-parameters of CC configuration are used. We may make use of Table 1 and simultaneously use conversion Table 2 to get the expression for  $A_i$ ,  $R_i$ ,  $A_v$  and  $R_o$ .



**Table 1: Result of small signal analysis of low frequency ce amplifier**

$A_I = -\frac{h_{fe}}{1 + h_{oe} \times R_L}$	$A_V = \frac{A_I \times R_L}{R_i}$
$R_i = h_{ie} + h_{re} \times A_I \times R_L$	$A_{Vs} \approx \frac{A_V R_i}{R_i + R_s}$
$Y_0 = h_{oe} - \frac{h_{re} h_{fe}}{h_{ie} + R_s} = \frac{1}{Z_0}$	$A_{IS} = \frac{A_I R_s}{R_i + R_s}$

**Table**

<i>From CB to CE</i>	<i>From CE to CB</i>	<i>From CE to CC</i>
$h_{ie} = \frac{h_{ib}}{1 + h_{\beta}}$	$h_{ie} = \frac{h_{ie}}{1 + h_{fe}}$	$h_{ic} = h_{ic}$
$h_{oe} = \frac{h_{ob}}{1 + h_{\beta}}$	$h_{ob} = \frac{h_{oe}}{1 + h_{fe}}$	$h_{oc} = h_{oe}$
$h_{fe} = \frac{h_{\beta}}{1 + h_{\beta}}$	$h_{\beta} = \frac{h_{fe}}{1 + h_{fe}}$	$h_{fe} = -(1 + h_{fe})$
$h_{re} = \frac{h_{ib} h_{ob}}{1 + h_{\beta}} \quad h_{rb}$	$h_{rb} = \frac{h_{ie} h_{oe}}{1 + h_{fe}} \quad h_{re}$	$h_{re} = 1 - h_{re} \cong 1$

Thus, from Table 1 and Table 2,

$$\text{Current gain } A_I = -\frac{I_e}{I_b} = -\frac{h_{fc}}{1 + h_{oc} \times R_L} = \frac{1 + h_{fe}}{1 + h_{oe} \times R_L} \quad \dots(1)$$

$$\text{Input resistance } R_i = \frac{V_i}{I_b} = h_{ic} + h_{re} \times A_I \times R_L = h_{ie} + A_I \times R_L \quad \dots(2)$$

$$\text{Voltage gain } A_V = \frac{V_e}{V_i} = \frac{A_I \times R_L}{R_i} \quad \dots(3)$$

But from equation (2)  $A_I \times R_L = R_i - h_{ie}$

$$\text{Hence } A_V = \frac{R_i - h_{ie}}{R_i} = 1 - \frac{h_{ie}}{R_i} \quad \dots(4)$$

$$\text{Overall voltage gain } A_{VS} = \frac{A_V \times R_i}{R_i + R_s} \quad \dots(5)$$

$$\text{Overall current gain } A_{IS} = \frac{A_I \times R_s}{R_i + R_s} \quad \dots(6)$$

$$\text{Output admittance } Y_0 = h_{oc} - \frac{h_{fc} \times h_{rc}}{h_{ic} + R_s} = h_{oe} + \frac{1 + h_{fe}}{h_{ie} + R_s} \quad \dots(7)$$

**For reference**

Table		
From CB to CE	From CE to CB	From CE to CC
$h_e \frac{h_{ib}}{1 + h_{fb}}$	$h_e \frac{h_{ie}}{1 + h_{fe}}$	$h_{ic} = h_{ie}$
$h_{oc} \frac{h_{ob}}{1 + h_{fb}}$	$h_{ob} \frac{h_{oc}}{1 + h_{fe}}$	$h_{oc} = h_{oe}$
$h_{fe} \frac{h_{fb}}{1 + h_{fb}}$	$h_{fb} \frac{h_{fe}}{1 + h_{fe}}$	$h_{fe} = -(1 + h_{fb})$
$h_{re} \frac{h_{fb} h_{ob}}{1 + h_{fb}} \quad h_{rb}$	$h_{rb} \frac{h_e h_{oc}}{1 + h_{fe}} \quad h_{re}$	$h_{re} = 1 - h_{re} \cong 1$



# Summary

$$A_I = -\frac{I_e}{I_b} = \frac{-h_{fc}}{1+h_{oc}R_L} = \frac{1+h_{fe}}{1+h_{oe}R_L} \quad \text{--- (1)} \quad \because h_{fc} = -(1+h_{fe}) \quad \text{(4)}$$

$$R_i = \frac{V_i}{I_b} = h_{ic} + h_{rc} A_I R_L = h_{ie} + A_I R_L \quad \text{--- (2)} \quad \because h_{rc} = h_{ie}$$

$$h_{rc} \approx 0$$

$$A_V = \frac{V_o}{V_i} = \frac{A_I R_L}{R_i} = \frac{R_i - h_{ie}}{R_i} = 1 - \frac{h_{ie}}{R_i} \quad \text{--- (3)} \quad \because A_I R_L = R_i - h_{ie} \text{ from eq (2)}$$

$$Y_o = h_{oc} - \frac{h_{fc} h_{rc}}{h_{ic} + R_s} = h_{oe} + \frac{1+h_{fe}}{h_{ie} + R_s} \quad \text{--- (4)} \quad \because h_{fc} = -(1+h_{fe})$$

$$h_{rc} \approx 1$$

$$h_{ic} = h_{ie}, \quad h_{oc} = h_{oe}$$

Ex 2: The emitter follower (CC amplifier) has the following h-parameters:

$h_{ie} = 1100 \Omega$ ,  $h_{re} = 2.5 \times 10^{-4}$ ,  $h_{fe} = 50$  and  $h_{oe} = 24 \mu A/V$ . If  $R_L = 10 k\Omega$  and  $R_S = 1 k\Omega$ , find the  $A_I$ ,  $R_i$ ,  $A_v$  and  $Y_o$ .

Solution:

$$A_I = \frac{1 + h_{fe}}{1 + h_{oe} R_L} = 40.75$$

$$R_i = h_{ie} + A_I R_L = 409 k\Omega$$

$$A_v = 1 - \frac{h_{ie}}{R_i} = 0.997$$

$$R_o = \frac{1}{Y_o} = \frac{1}{(h_{oe} + \frac{1 + h_{fe}}{h_{ie} + R_S})} = 41.2 \Omega$$

$$A_{vS} = \frac{A_v R_i}{R_i + R_S} = 0.997 \times \frac{409}{410} = 0.995$$